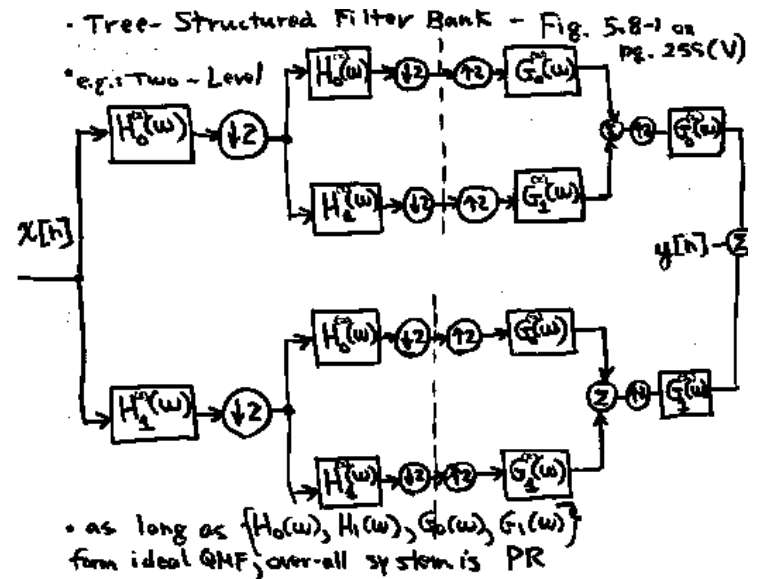


EE648 (CC761-M) DSP II

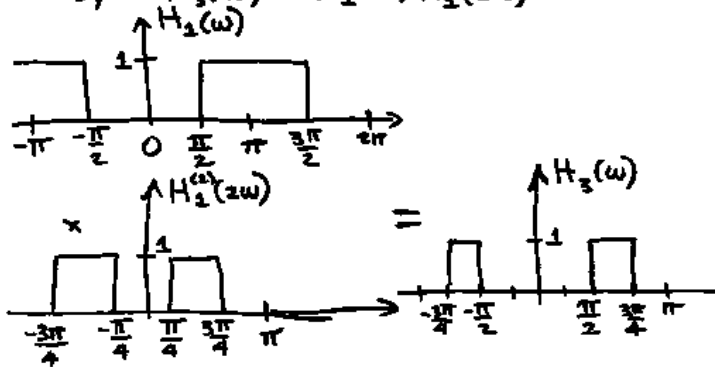
Session 22 (live: 4/1/99)

Outline

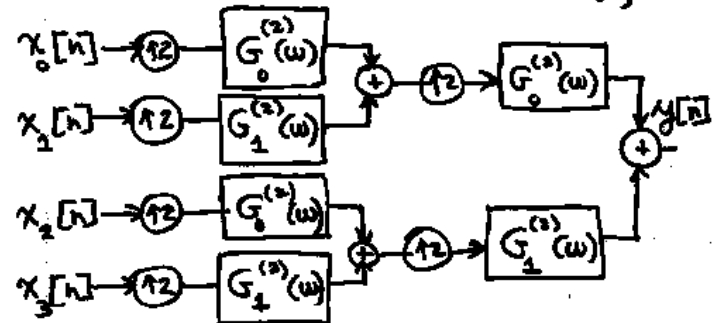
- Synthesis of M-channel uniform filter banks from tree-structured filter banks - Sect. 5.8 of Y.
- Analysis of M-channel PR filter banks - Sect. 5.4 of Y.



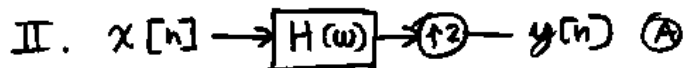
• What frequency band is passed by $H_3(w) = H_2^{(1)}(w) H_2^{(2)}(2w)$?



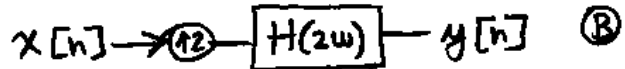
• Synthesis section of tree-structured filter bank e.g. $M=2^2$



• to show equivalence to $M=4$ channel filter bank, use Noble's second identity



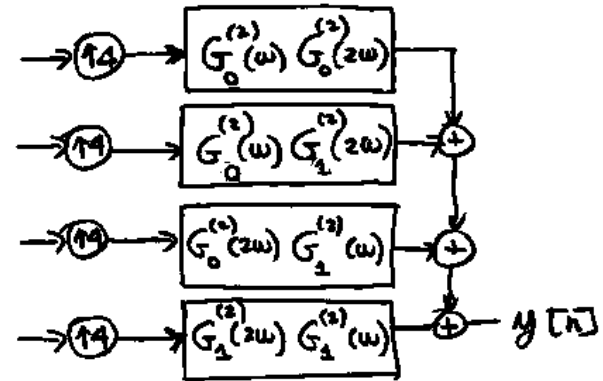
• same I/O relationship with



(A) $Y(\omega) = H(2\omega) X(2\omega)$
 (B) $Y(\omega) = H(2\omega) X(2\omega)$ } SAME I/O

• See PR 4 chan. m (based on $h[n] = \{1, 1\}$) and PR RC 4 chan. m (based on $h[n]$ having a raised cosine spectrum) at course web site

• equivalent $M=4$ uniform filter bank:



• Analysis of M -channel filter bank

$$X_m(\omega) = \sum_{l=0}^{M-1} H_m\left(\frac{\omega - l2\pi}{M}\right) X\left(\frac{\omega - l2\pi}{M}\right)$$

$$W_m(\omega) = X_m(M\omega) = \sum_{l=0}^{M-1} H_m\left(\omega - l\frac{2\pi}{M}\right) X\left(\omega - l\frac{2\pi}{M}\right)$$

$$Y(\omega) = \sum_{m=0}^{M-1} G_m(\omega) W_m(\omega) = \sum_{m=0}^{M-1} G_m(\omega) \sum_{l=0}^{M-1} H_m\left(\omega - l\frac{2\pi}{M}\right) X\left(\omega - l\frac{2\pi}{M}\right)$$

$$Y(\omega) = \sum_{l=0}^{M-1} X(\omega - l \frac{2\pi}{M}) \sum_{m=0}^{M-1} G_m(\omega) H_m(\omega - l \frac{2\pi}{M})$$

define: $F_l(\omega) = \sum_{m=0}^{M-1} G_m(\omega) H_m(\omega - l \frac{2\pi}{M})$

$$Y(\omega) = F_0(\omega) X(\omega)$$

$$+ \sum_{l=1}^{M-1} F_l(\omega) X(\omega - l \frac{2\pi}{M})$$

• desire: $|F_0(\omega)| = \text{constant} \geq \forall \omega$
 $\angle F_0(\omega) = \text{linear}$

$$F_l(\omega) = 0 \quad \forall \omega$$

for $l=1, \dots, M-1$