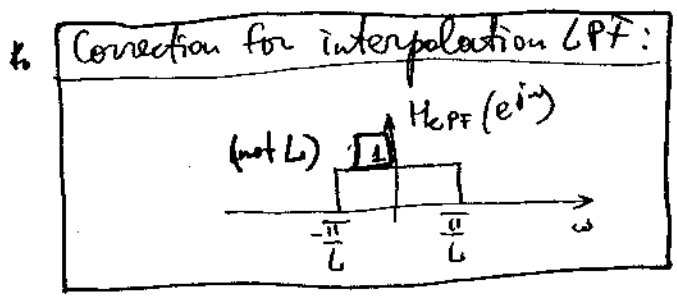


Lec 12, Wed 9/19/2001 ①



1.5 Z-transform.

Last time, we started studying the Z-transform, which, as you are seeing in this week's lab, is an important tool, in particular, for filter design.

~~1.5.1 Rational Z-Transforms.~~

1.5.1. Rational Z-Transforms.

$$y(n) = \sum_{i=0}^{N-1} b_i x(n-i) - \sum_{k=1}^M a_k y(n-k)$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{i=0}^{N-1} b_i z^{-i}}{1 + \sum_{k=1}^M a_k z^{-k}} = b_0 z^{M-N+1} \frac{\prod_{i=0}^{N-1} (z - z_i)}{\prod_{k=1}^M (z - p_k)}$$

← zeros
 ← poles

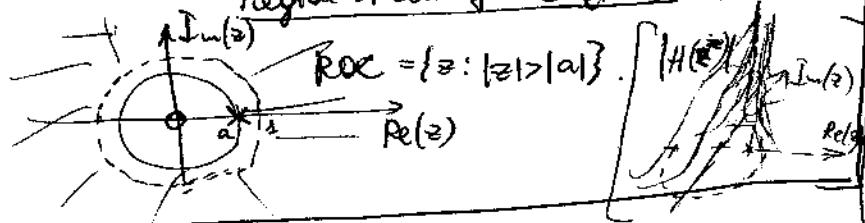
if $b_0 \neq 0$

Values of z for which $X(z)=0$ are called the zeros of $X(z)$:
 z_1, \dots, z_{N-1} - roots of the numerator
 Values of z for which $X(z)$ is infinite are called the poles of $X(z)$:
 p_1, \dots, p_M - roots of the denominator
 In addition, $M-N+1$ zeros at $z=0$ if $M > N-1$, or
 $N-1-M$ poles at $z=0$ if $N-1 > M$
 Poles and zeros may also occur at $z=\infty$:
 "zero at ∞ " means $\lim_{|z| \rightarrow \infty} X(z) = 0$
 "pole at ∞ " means $\lim_{|z| \rightarrow \infty} X(z) = \infty$.

Answer

Example. $ZT \{a^n u(n)\} = \sum_{n=0}^{\infty} a^n z^{-n} = \sum_{n=0}^{\infty} (az^{-1})^n = \frac{1}{1-az^{-1}} = \frac{z}{z-a}$ if $|az^{-1}| < 1$
 region of convergence (ROC): $|z| > |a|$

Pole: $z=a$
 zero: $z=0$

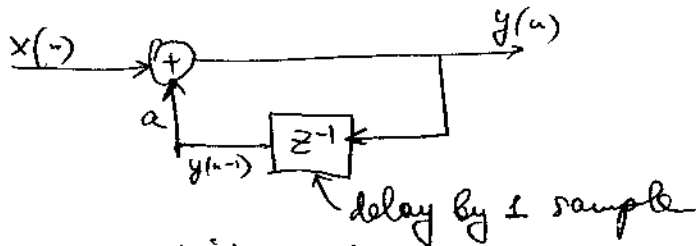


Let's analyze the filter whose transfer function is this

Let $H(z) = \frac{z}{z-a}$

(a) Diff. eq.:
 $y(n) = ay(n-1) + x(n)$ Recursive IIR

(b) Diagram:

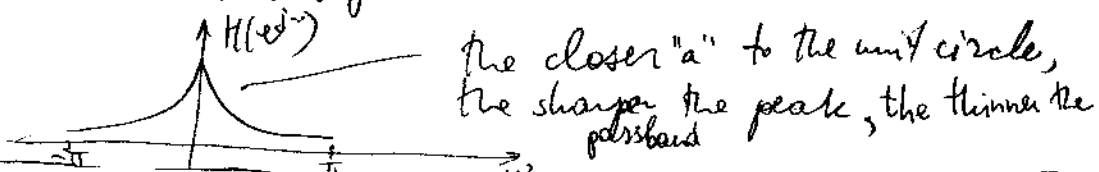


(c) Freq. resp: $H(e^{j\omega}) = \frac{1}{1-ae^{-j\omega}}$

$$|H(e^{j\omega})| = \left| \frac{1}{1-ae^{-j\omega}} \right| = \frac{1}{\sqrt{(1-ae^{-j\omega})(1-ae^{j\omega})}} = \frac{1}{\sqrt{1-2a\cos\omega+a^2}}$$

$$= \frac{1}{\sqrt{(1-a)^2 + 2a(1-\cos\omega)}}$$

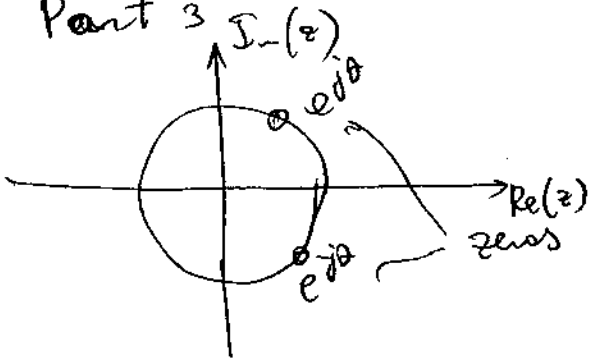
$a=0.99 \Rightarrow$ denominator is very small for $\omega=0$:



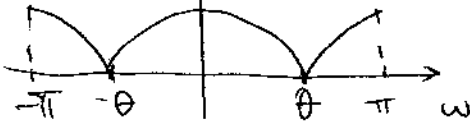
- In general,
- a pole near the unit circle will cause the freq. resp. to ~~be~~ increase (5)
~~decrease~~ in the width of that pole
 - a zero near the unit circle will cause the freq. resp. to ~~be~~ decrease
~~increase~~ in the width of that zero.

Lab 5 wk 4

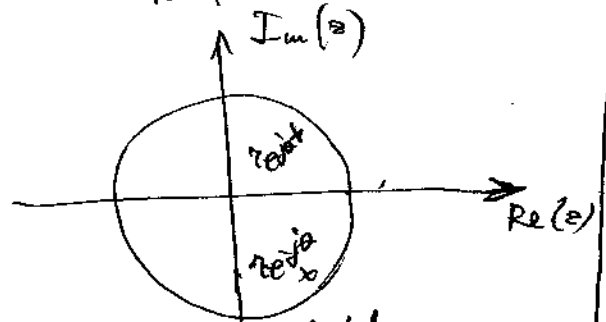
Part 3



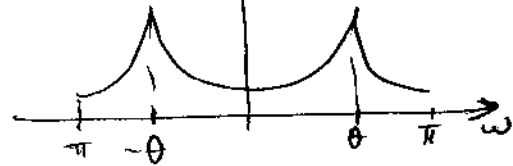
bandstop filter
 $|H(e^{j\omega})|$



Part 4



bandpass filter.
 $|H(e^{j\omega})|$



1.5.2 Region of Convergence (ROC) of the z-Transform.

Example 1. $x(n) = 2^n u(n)$

$$X(z) = \sum_{n=0}^{\infty} 2^n z^{-n} = \sum_{n=0}^{\infty} \left(\frac{2}{z}\right)^n$$

When can we sum this series? (I.e., when does this series converge?)

(a) If $|z| \leq 2$, then $\left|\frac{2}{z}\right| \geq 1$
 \Rightarrow every term has abs. value ≥ 1 and growing (or not constant)
 \Rightarrow can't sum

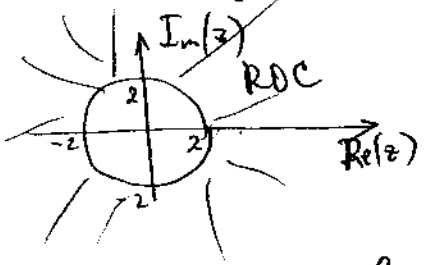


(b) If $|z| > 2$, then $\left|\frac{2}{z}\right| < 1$

$$\Rightarrow X(z) = \frac{1}{1 - \frac{2}{z}}$$

Thus, the z-transform of $x(n) = 2^n u(n)$ is

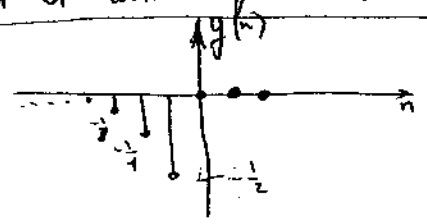
$$X(z) = \begin{cases} \frac{1}{1 - \frac{2}{z}}, & |z| > 2 \\ \text{undefined}, & |z| \leq 2 \end{cases}$$



Usually called "the region of convergence (ROC) of the z-transform", because it's where this series actually converges to this function. A slightly better term would be "the region of definition", since outside this region, the z-transform of this sequence is simply undefined.

Example 2. $y(n] = -2^n u(-n-1)$

$$Y(z) = \sum_{n=-\infty}^{-1} -2^n z^{-n} = -\sum_{m=0}^{\infty} 2^{-m-1} z^{m+1} = -\frac{z}{2} \sum_{m=0}^{\infty} \left(\frac{z}{2}\right)^m$$



(a) $|z| \geq 2 \Rightarrow$ can't sum

(b) $|z| < 2 \Rightarrow Y(z) = -\frac{z}{2} \frac{1}{1 - \frac{z}{2}} = -\frac{z}{2-z} = \frac{1}{1 - \frac{2}{z}}$

$$Y(z) = \begin{cases} \text{undefined}, & |z| \geq 2 \\ \frac{1}{1 - \frac{2}{z}}, & |z| < 2 \end{cases}$$

The same expression for the z-transform as in Example 1, but non-intersecting regions of convergence.

Example 3. $w(n) = 2^n$, $-\infty < n < \infty$.

$$w(n) = 2^n u(n) + 2^n u(-n-1) = x(n) - y(n)$$

$$W(z) = X(z) - Y(z) = \frac{1}{1 - \frac{z}{2}} - \frac{1}{1 - \frac{z}{2}} \stackrel{!}{=} 0 ?$$

What's wrong with this?

$X(z)$ and $Y(z)$ have no common ROC:

$X(z)$ is undefined for $|z| \leq 2$

$Y(z)$ " " " $|z| \leq 2$

$\Rightarrow W(z)$ is never defined.

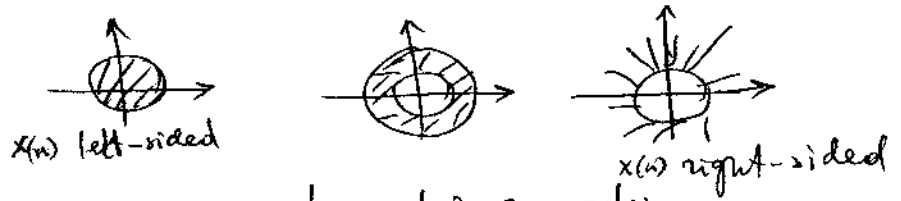
Def. ROC of $\sum_{n=-\infty}^{\infty} x(n)z^{-n}$ is the set of all z for which this series is absolutely convergent:

$$\sum_{n=-\infty}^{\infty} |x(n)z^{-n}| < \infty$$

(We use absolute convergence to avoid certain pathological series which converge, but do not converge absolutely.)

1.5.3. Properties of ROC, Poles, and Zeros.

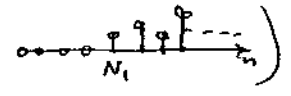
1) The ROC is a ring or a disc centered at the origin:
 $r_1 < |z| < r_2$ (r_1 and/or r_2 could be 0 or ∞)



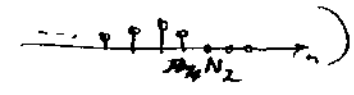
2) The ROC cannot contain any poles.

3) If $x(n)$ is a finite-duration sequence (i.e., $x(n) = 0$ except for $N_1 \leq n \leq N_2$), then ROC is the whole z -plane except possibly $z = 0$.

4) If $x(n)$ is a right-sided sequence ($x(n) = 0$ for $n < N_1$), then ROC is $|z| > |p_{\max}|$



5) If $x(n)$ is left-sided ($x(n) = 0$ for $n > N_2$), then ROC is $|z| < |p_{\min}|$



innermost non-zero pole in $X(z)$

6) An LTI system is BIBO stable if and only if the ROC of its transfer function $H(z)$ includes $z = 1$ (i.e., includes the unit circle):

$$\left[\sum_n |h(n) z^{-n}| \right]_{z=1} < \infty$$

$$\sum_n |h(n)| < \infty$$

BIBO stable

(Again, ONLY if LTI)