

AR model:

$$S(n) = GX(n) + \sum_{k=1}^p a_k S(n-k)$$

Why would such a model be useful?

Synthesis:

model for each phoneme → synthesize speech
(toy example in Lab 9, wk 1)

Coding:

1. Identify parameters over short time intervals.
2. Transmit/store a stream of model parameters: one set per interval, + the error signal.
3. Reconstruct. ~~parameters~~

Recognition

- store a model for every sound (e.g., every phoneme).
- Given a speech signal to be identified, estimate "the best" model parameters.
- match them with the stored ones.

(Also, note that such ^{and similar} models are used in many other areas besides speech processing (e.g., finance, image processing, etc.))

Question: Given speech signal (s), how do we generate this model? Need to estimate the following parameters:

1. Voiced/unvoiced (we talked a little about this last time)
2. DT pitch period (for now, assume we know this)
3. Order p (will also assume fixed and given. can get reasonable models with p around 15.)
4. Gain G
5. Filter coefficients a_1, \dots, a_p
 ↙ similar; we assume $G=1$ and try to estimate a_1, \dots, a_p
6. Delay m. (will model as random).

2.1.1 Parameter Estimation for the AR Model.

Observe a voiced sound $S(n)$; want to estimate the corresponding a_1, \dots, a_p .

Problem: we don't know when $X(n)$ started - i.e., don't know m .
 One possible solution: model m as ~~an~~ random between 0 and $D-1$, i.e., the input to the system is a random sequence:

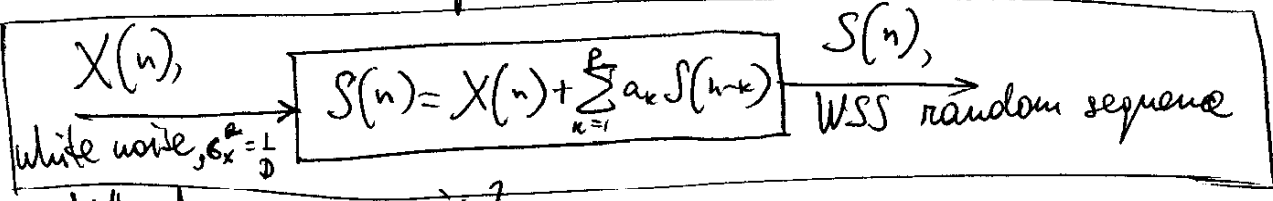
$$X(n) = x^{(m)}(n) \text{ with probability } \frac{1}{D}, m=0, 1, \dots, D-1,$$

where ~~$x^{(m)}(n) = \sum_{k=-\infty}^{\infty} \delta(n-Dk-m)$~~ $x^{(m)}(n) = \sum_{k=-\infty}^{\infty} \delta(n-Dk-m) =$

$$E[(X(n))^2] = \sum_{m=0}^{D-1} (x^{(m)}(n))^2 \frac{1}{D} = \frac{1}{D}$$

$$E[X(n)X(n+l)] = \begin{cases} 0, & l \text{ is not a multiple of } D \\ \frac{1}{D}, & \text{otherwise} \end{cases}$$

I.e., $X(n)$ is white over $[0, D-1]$.
 We will make a further approximation and model $X(n)$ as a white noise process with variance $\frac{1}{D} = \sigma_x^2$.



What are a_k 's?
 Multiply both sides by $S(n-l)$, and take expected values:

(4)

$$E[S(n)S(n-l)] = E[X(n)S(n-l)] + \sum_{k=1}^p a_k E[S(n-k)S(n-l)]$$

$$r_{ss}(l) = \underbrace{c_{xs}(-l)}_{\substack{\text{cross-correlation} \\ \text{function of } X \text{ and } S}} + \sum_{k=1}^p a_k r_{ss}(l-k)$$

(1)

autocorrelation
fun of S

cross-correlation
function of X and S

Recall that $c_{xs}(-l) = E[X(n)S(n-l)]$
 $= E\left[\underbrace{\sum_{k=-\infty}^{\infty} h(k) X(n-l-k)}_{S(n-l)} X(n)\right]$

$$= \sum_{k=-\infty}^{\infty} h(k) r_{xx}(k+l)$$

since X is
white \Rightarrow

$$\sum_{k=-\infty}^{\infty} h(k) \sigma_x^2 \delta(k+l) = \sigma_x^2 h(-l)$$

since the system is causal,
and since $h(0)=1$

$$= \begin{cases} 0 & l > 0 \\ \sigma_x^2 & l = 0 \\ \sigma_x^2 h(l) & l < 0 \end{cases}$$

$S(n)$ is uncorrelated
with future values of $X(n)$
(2)

Putting (1) and (2) together,

$$r_{ss}(l) = \sum_{k=1}^p a_k r_{ss}(l-k), \quad l > 0$$

"Yule-Walker equations"

(Also note that

$$r_{ss}(0) = \sum_{k=1}^p a_k r_{ss}(0-k) + \sigma_x^2$$

$$= \sum_{k=1}^p a_k r_{ss}(k) + \sigma_x^2;$$

and $r_{ss}(l) = r_{ss}(-l); \quad l < 0.$

(5)

In matrix form,

$$\begin{pmatrix} r_{ss}(1) \\ r_{ss}(2) \\ \vdots \\ r_{ss}(p) \end{pmatrix} = \begin{pmatrix} r_{ss}(0) & r_{ss}(-1) & \dots & r_{ss}(-p+1) \\ r_{ss}(1) & r_{ss}(0) & \dots & r_{ss}(-p+2) \\ \vdots & \vdots & \ddots & \vdots \\ r_{ss}(p-1) & r_{ss}(p-2) & \dots & r_{ss}(0) \end{pmatrix} \begin{pmatrix} a_1 \\ a_2 \\ \vdots \\ a_p \end{pmatrix}$$

- Estimate the autocorrelation of $S(n)$ (recall Lab 7)
- Estimate a_1, \dots, a_p by solving Yule-Walker equations - i.e. by inverting the autocorrelation matrix
- To synthesise speech using the estimated a_k 's, we can excite the system with $\sum_{k=-\infty}^{\infty} \delta(n-D_k)$
- see Lab 9 wk 1, Sect 2.4

(I.e., for system identification we assumed $X(n)$ to be white noise since we didn't know $X(n)$ exactly. But for synthesis we can use a more correct $\sum_{k=-\infty}^{\infty} \delta(n-D_k)$.)

2.2 Linear Prediction.

(6)

Suppose we observe a WSS random sequence $S(n)$ and would like to predict the next sample as a linear combination of q previous samples:

$$\hat{S}(n) = \sum_{k=1}^q d_k S(n-k) \quad (*)$$

Prediction error: $E(n) = S(n) - \hat{S}(n)$.

Find the predictor coefficients so as to minimize the mean square error, $e = E[(E(n))^2]$.

$$\begin{aligned} \frac{\partial e}{\partial d_l} &= \frac{\partial}{\partial d_l} E \left\{ \left(S(n) - \sum_{k=1}^q d_k S(n-k) \right)^2 \right\} \\ &= E \left\{ 2 \left(S(n) - \sum_{k=1}^q d_k S(n-k) \right) (-S(n-l)) \right\} \\ &= -2 r_{SS}(l) + 2 \sum_{k=1}^q d_k r_{SS}(l-k) \end{aligned}$$

Set this to zero:

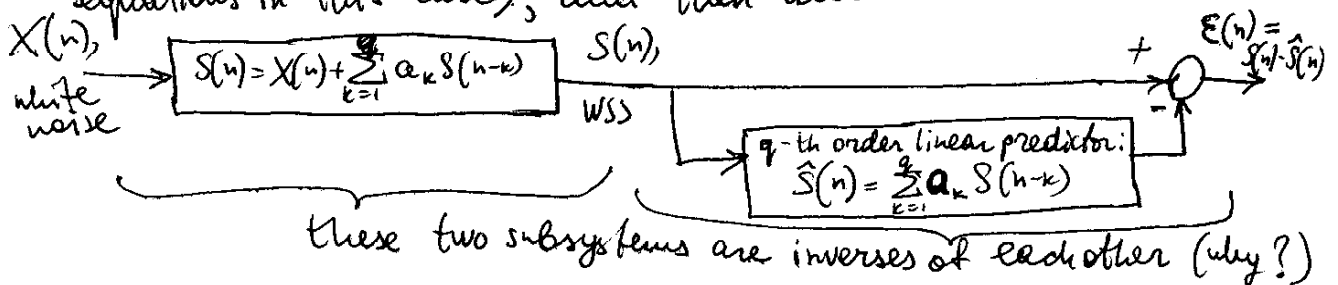
$$r_{SS}(l) = \sum_{k=1}^q d_k r_{SS}(l-k), \quad l=1, \dots, p$$

- normal equations

Remarks:

1. $\frac{\partial^2 e}{\partial d_l^2} = E[2(S(n-l))^2] > 0$ unless $S(n) \equiv 0$
 \Rightarrow what we found is a minimum.

2. If $S(n)$ is a q -th order AR process, then $d_k = a_k$ (the normal equations are identical to the Yule-Walker equations in this case), and then also



Then $\varepsilon(n) = X(n) =$ white noise.

(7)

The linear predictor acts as a "whitening filter", i.e., removes correlation from $S(n)$, producing a white error process.

3. Let's step back and think what exactly we are doing here. ~~We are~~ This prediction problem is a linear least-squares estimation problem:

we are trying to estimate random variable $S(n)$ as a linear combination of the random variables

$S(n-1), \dots, S(n-p)$, by minimizing the mean-square error.