RTN and Low Frequency Noise on Ultra-scaled Near-ballistic Ge Nanowire nMOSFETs

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Abstract

In this work, we present the first observation of random telegraph noise (RTN) in ultra-scaled Ge nanowire (NW) nMOSFETs. The impacts of NW geometry, channel length, EOT, and channel doping on low frequency noise are studied comprehensively. It is confirmed that the low frequency noise with 1/f characteristics is attributed to the mobility fluctuation in ultra-scaled Ge NW nMOSFETs. The low frequency noise decreases when the channel length scales down from 80 nm to 40 nm because of the near-ballistic transport of electrons.

Introduction

Ge channel is one of the promising candidates for future ultimate CMOS applications because Ge has high and balanced electron and hole mobility, good compatibility with Si large-scale-integration technologies and great potential for voltage scaling [1-2]. Ge CMOS has been intensively studied in the past decade and highly scaled Ge NW CMOS has been demonstrated to offer excellent performance [3-5]. However, advanced high-k/Ge gate stacks with scaled EOT and superior MOS interfaces are still needed to develop Ge CMOS manufacturing technology with high reliability [6]. The conventional C-V method and charge pumping method cannot be applied to ultrasmall devices without a body contact. It is increasingly hard to directly measure the interface and oxide property when the devices are aggressively scaled down to sub-100 nm. Therefore, low frequency noise and RTN can be used as alternate probes for device characterization and process optimization since noise measurement is not limited by the small gate capacitance [7-9]. In the meanwhile, low frequency noise and RTN have serious impacts in scaled nonvolatile memories and logic circuits [10-11]. Several groups have reported low frequency noise study of long channel Ge MOSFETs [12-13]. There is still no study on low frequency noise and RTN of highly scaled Ge MOSFETs with sub-100 nm gate length.

In this paper, we: 1) report the first observation of RTN in ultrascaled Ge NW nMOSFETs and extract the basic parameters of the traps; 2) examine the origin of the 1/f low frequency noise; 3) systematically study the properties of low frequency noise on Ge NW nMOSFETs with various NW geometries (NW width, W_{NW} , NW height, H_{NW}), channel length (L_{ch}), EOT and channel doping.

Experiments

The fabrication process is shown in Fig. 1, which is the same as reported in Ref. [5]. Fig. 2 illustrates the schematic image of the Ge NW nMOSFETs. The cross-section view shows the accumulation mode (AM), inversion mode (IM) nMOSFETs and the key geometry parameters of devices. Ge NW nMOSFETs with W_{NW} from 10 nm to 40 nm, H_{NW} of 7 nm and 10 nm, L_{ch} from 40 nm to 80 nm and EOT of 2 nm and 5 nm are used for RTN and low frequency noise characteristics. The channel width is calculated from $W_{ch} = (2 \cdot H_{NW} + W_{NW}) \times (number of wires)$ and each device has 7 nanowires. AM NW nMOSFETs are measured unless otherwise specified. Keysight B1500A with B1530A Waveform Generator/Fast Measurement Unit is applied for characterization. All the measurements were performed at room temperature at a drain voltage (V_{ds}) of 50 mV.

Results and Discussion

<u>A. RTN characteristics</u>: Figs. 3-4 show the good output and transfer characteristics of a Ge NW nMOSFET with L_{ch} of 40 nm and EOT of 2 nm. Typical RTN signal of drain current (I_d) with signal trap and two traps are illustrated in Fig. 5. Two and four distinct current switching levels are observed in the two devices with the same device dimension. The trap number *n* can be estimated from current fluctuation levels N via $2^{n-1} < N \le 2^n$. The corresponding histograms of I_d with two and four peaks are shown in Fig. 6. The histograms show the superposition of RTN signal and mobility fluctuation (1/f) noise. Average time constants in single trap RTN,

such as capture time constant (τ_c) and emission time constant (τ_c), are extracted by exponential fitting to time constants distributions (Fig. 7). τ_c and τ_e may change with gate voltage (V_{gs}) because of the move of Fermi level. Clear RTN signals were observed when V_{gs} was near threshold voltage (V_{th}). The relation between τ_c , τ_e , τ_c/τ_e and V_{gs} is shown in Fig. 8. The negative correlation between τ_c/τ_e and V_{gs} indicates that the electrons' trapping and de-trapping occur between channel and gate oxide [9]. The depth of this trap from the interface (x_T) in the gate oxide can be calculated according to the formula in Fig. 8. For the trap in 1# device of Fig. 5, x_T equals to 0.6T_{OX} (oxide thickness), indicating that it is a deep trap.

<u>B. Low frequency noise:</u> Fig. 9 shows the power spectrum density (PSD) of I_d (S_{Id}) in 1# device of Fig. 5 under different gate voltages. Typical Lorentzian spectrum with 1/f² characteristics is observed. The low frequency noise increases with higher V_{gs}. The normalized S_{Id} (S_{Id}/I_d^2) between two devices, sharing the same device dimension, with and without RTN signal is compared in Fig. 10. The noise spectrum of device without RTN shows 1/f characteristics. The low frequency noise results (S $_{\rm Id},~S_{\rm Vg})$ in Figs. 11-16 are obtained in devices without RTN at a frequency of 10 Hz and each group of the data contains experimental results from several devices sharing the same device dimension. Fig. 11 shows the normalized $S_{Id} (S_{Id}/I_d^2)$ as a function of Id. The clear 1/Id dependence agrees well with the mobility fluctuation model [14]. Fig. 12 presents the input gate noise $(S_{Vg}=S_{Id}/g_m^2)$ normalized by channel area $(W_{ch}\cdot L_{ch}\cdot S_{Vg})$ versus I_d . The normalized S_{Vg} decreases when the L_{ch} scales down, while the classical theory indicates $W_{ch}{\cdot}L_{ch}{\cdot}S_{Vg}$ shall be independent of channel area [14]. The anomalous scaling trend of low frequency noise can be attributed to the near-ballistic transport of electrons in the channel. Because of the electron's long mean free path in Ge, electrons encounter less scattering at smaller Lch. Therefore, the scattering induced mobility fluctuation decreases at small L_{ch}, thus the normalized S_{Vg} reduces. Similar results were also observed in highly scaled InGaAs MOSFETs as reported in Ref. [15]. Also, the parabolic increase of S_{Vg} confirms again the low frequency noise is attributed to the mobility fluctuation other than the carrier number fluctuation [16]. Since scattering is an important source of low frequency noise, S_{Vg} is smaller in IM NW MOSFETs because the relative lower channel doping would induce less Coulomb scattering (Fig. 13). S_{Vg} decreases in device with smaller EOT because of the enhancement of gate control (Fig. 14). The change in NW geometries affects the low frequency noise due to the change of channel area. Since low frequency noise is inversely proportional to the channel area, S_{Vg} increases with the decrease in W_{NW} and H_{NW} (Figs. 15-16). In the perspective of device performance, the increase of H_{NW} promises the enhancement of on-state performance, i.e., I_d and g_m per pitch area, and the suppression of low frequency noise.

Conclusion

We report the first observation of RTN in Ge NW nMOSFETs. The mobility fluctuation is confirmed to be the source of low frequency noise in ultra-scaled Ge NW nMOSFETs other than the carrier number fluctuation. Because of the long mean free path of electrons in Ge, the low frequency noise is suppressed at shorter channel due to the near-ballistic transport at sub-100 nm region. Therefore, ultra-scaled Ge NW MOSFETs with low channel doping, small EOT and high $H_{\rm NW}$ are promising in the performance enhancement as well as the suppression of low frequency noise.

Reference

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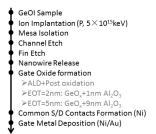


Fig. 1 Fabrication process flow of the Ge NW nMOSFETs.

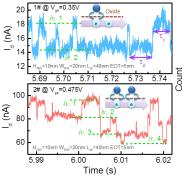


Fig. 5 I_d fluctuation due to RTN with single trap (upper) and two traps different (lower) in two Ğe nMOSFETs having the same device dimension (1#, 2#).

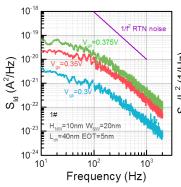


Fig. 9 S_{Id} of RTN signal in 1# device of Fig. 5 at different gate voltages, showing 1/f² characteristics.

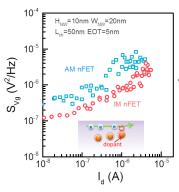


Fig. 13 S_{Vg} versus I_d in AM and IM nMOSFETs. S_{Vg} is smaller in IM nMOSFETs because of the less Coulomb scattering in the channel.

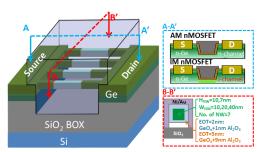


Fig. 2 Schematic diagram of the Ge NW nMOSFETs. The cross-section view shows the accumulation mode (AM), inversion mode (IM) nMOSFETs and the key geometry parameters.

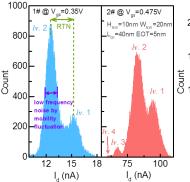


Fig. 6 Histograms of the RTN signal in 1# and 2# devices shown in Fig. 5. Two and four peaks are observed respectively.

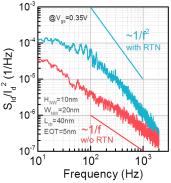


Fig. 10 Normalized $S_{Id}\ of\ two$ devices, having the same device dimension, with and w/o RTN. $S_{Id}\xspace$ in device w/o RTN shows 1/fcharacteristics.

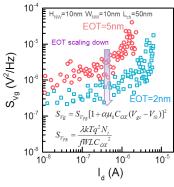
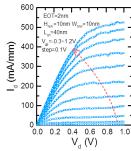


Fig. 14 S_{Vg} versus I_d in devices with different EOT. S_{Vg} decreases with the scaling down of EOT because of the enhanced gate control.



I....=10nm 10² =10nn 10¹ I_a (mA/mm) 100 10 DIBL=106mV/V SS(0.5V)=91mV/dec 10-2 10 0.9 1.2

Fig. 4 Transfer characteristics of the same device in Fig. 3. SS of 91 mV/dec is achieved

q

 $X_T = 0.6 \cdot T_{OX}$

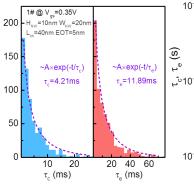
0.36

 ∂V_g

0

10-

10-2



nm.

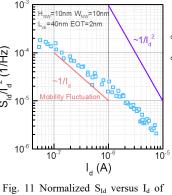
and τ_e are calculated by fitting.

to different gate voltages in 1# device of Fig. 5. The trap depth (x_T) can be obtained.

 $V_{gs}(V)$

=10nm W....=20nm =40nm EOT=5nm

0.32



devices w/o RTN. 1/I_d dependence

fluctuation

mobility

indicates

induced noise.

H_{NW}=10nm W_M=10nn 40nm 10 50nm W_{ch}·L_{ch}·S_{Vg} (μm²·μV²/Hz) 60nm Δ EOT=2nm 80nm . ^ aling dow 10 10 10^{2} 10 10 10 10 $I_{d}(A)$

Fig. 12 $S_{\rm Vg}$ normalized by $W_{ch}\!\!\times\!\!L_{ch}$ of devices (w/o RTN) with different values of L_{ch}. Normalized S_{Vg} decreases with scaling down of L_{ch} .

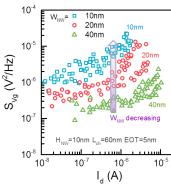


Fig. 15 S_{Vg} versus I_d in devices with different \tilde{W}_{NW} . S_{Vg} increases with the scaling down of W_{NW} because of the decrease of the gate areas.

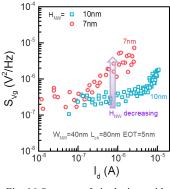
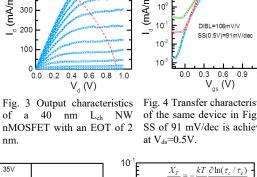


Fig. 16 S_{Vg} versus I_d in devices with different H_{NW} . S_{Vg} increases with the decrease in H_{NW} because of the decrease of the gate areas.

Fig. 7 Distribution of capture and Fig. 8 τ_c , τ_e and τ_c/τ_e corresponding emission time (τ_c, τ_e) constants of RTN signal in 1# device of Fig. 5. τ_c

10 S_{ld}/I_d² (1/Hz)



 T_{ox}

10